



AME

AME5259A

## 1.2A, 1.5MHz Synchronous Step-Down Converter

### ■ General Description

The AME5259A is a high efficiency monolithic synchronous buck regulator using a constant frequency, current mode architecture. Capable of delivering 1.2A output current over a wide input voltage range from 2.5V to 5.5V.

Supply current with no load is 400 $\mu$ A and drops to <1 $\mu$ A in shutdown. The 2.5V to 5.5V input Voltage range makes the AME5259A ideally suited for single Li-Ion battery-powered applications. 100% duty cycle provides low dropout operation, extending battery life in portable systems. PWM pulse skipping mode operation provides very low output ripple voltage for noise sensitive applications. At very light load, the AME5259A will automatically skip pulses in pulse skip mode operation to maintain output regulation.

The internal synchronous switch increases efficiency and eliminates the need for an external Schottky diode. Low output voltages are easily supported with the 0.6V feedback reference voltage. The AME5259A is available in small DFN-6D, QFN-16C and SOT-25 packages.

Other features include soft start, lower internal reference voltage with 2% accuracy, over temperature protection, and over current protection.

### ■ Features

- High Efficiency: Up to 95%
- Shutdown Mode Draws < 1 $\mu$ A Supply Current
- 2.5V to 5.5V Input Range
- Adjustable Output From 0.6V to  $V_{IN}$
- 1.0V, 1.2V, 1.5V, 1.6V, 1.8V, 2.5V and 3.3V Fixed/Adjustable Output Voltage
- 1.2A Output Current
- Low dropout Operation: 100% Duty Cycle
- No Schottky Diode Required
- 1.5MHz Constant Frequency PWM Operation
- Green Product Meet RoHS Standard

### ■ Applications

- Cellular Telephones
- Personal Information Appliances
- Wireless and DSL Modems
- MP3 Players
- Portable Instruments

### ■ Typical Application

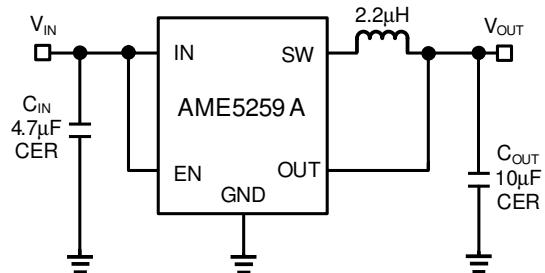
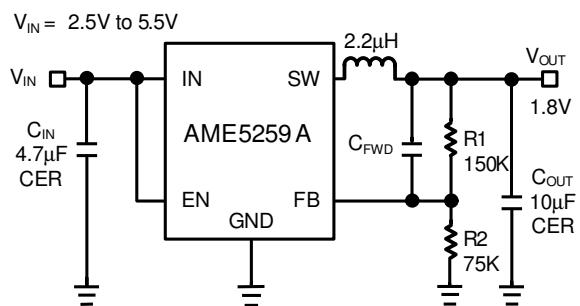


Figure 1. Fixed Output Voltage  
High Efficiency Step-Down Converte



$V_{IN} = 2.5V \text{ to } 5.5V$   
 $V_{OUT} = 1.8V$   
 $C_{FWD}: 22pF \text{ to } 220pF$   
Figure 2. Adjustable Output Voltage  
1.8V at 1000mA Step-Down Regulator

## ■ Functional Block Diagram

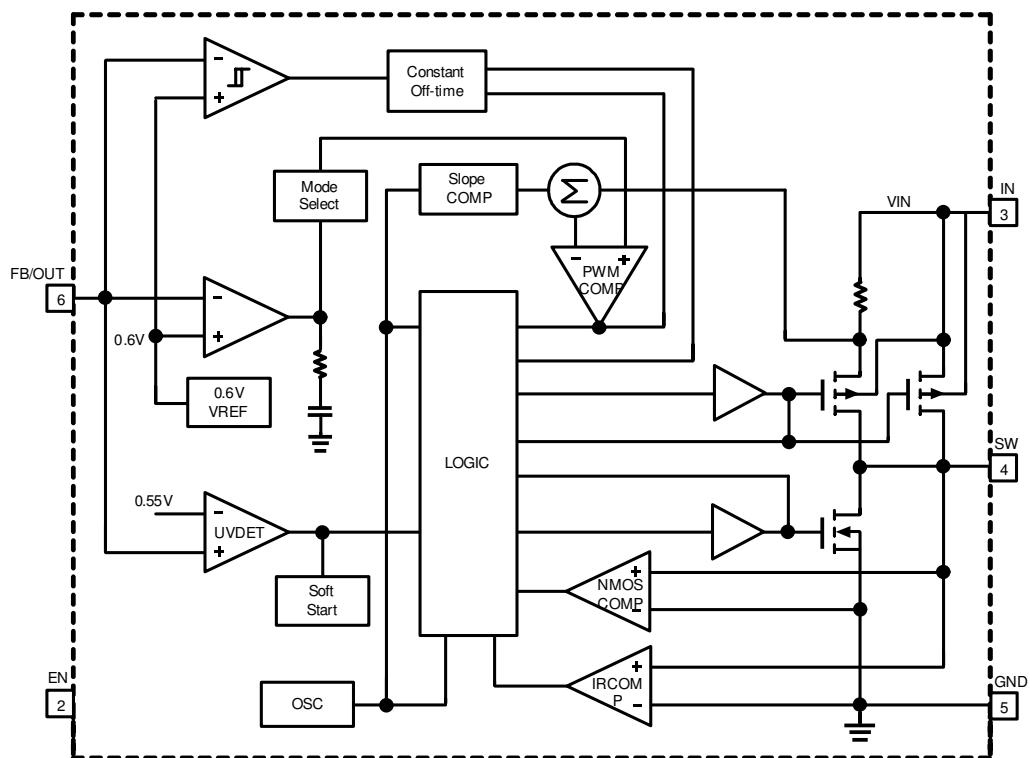


Figure 3. Function Block Diagram



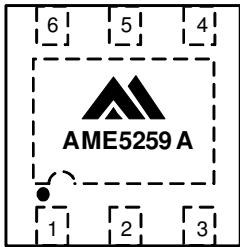
AME

AME5259A

1.2A, 1.5MHz Synchronous  
Step-Down Converter

## ■ Pin Configuration

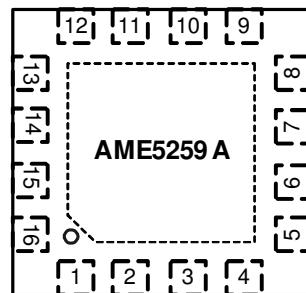
DFN-6D  
(2mmx2mmx0.75mm)  
Top View



AME5259A-AVYxxx

1. NC
2. EN
3. IN
4. SW
5. GND
6. FB/OUT

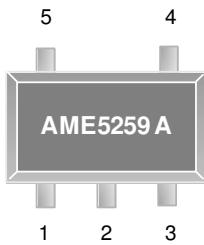
QFN-16C  
(3mmx3mmx0.75mm)  
Top View



AME5259A-AWExxx

- |           |        |
|-----------|--------|
| 1. GND    | 9. IN  |
| 2. GND    | 10. IN |
| 3. GND    | 11. IN |
| 4. FB/OUT | 12. IN |
| 5. GND    | 13. SW |
| 6. NC     | 14. SW |
| 7. EN     | 15. SW |
| 8. NC     | 16. NC |

SOT-25  
Top View



AME5259A-AEVADJ

1. EN
2. GND
3. SW
4. IN
5. FB/OUT

\* Die Attach:  
Conductive Epoxy

Note:

Connect exposed pad (heat sink on the back) to GND.



**AME5259A**

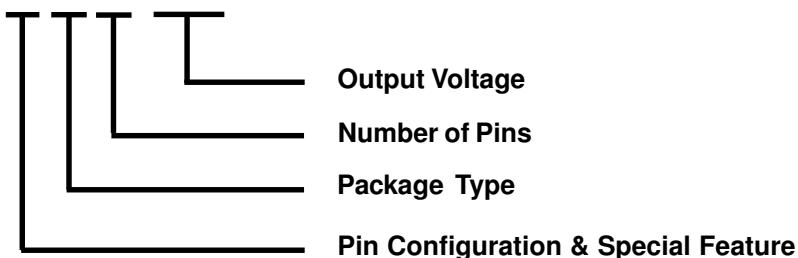
**1.2A, 1.5MHz Synchronous  
Step-Down Converter**

## ■ Pin Description

Pin Number			Pin Name	Pin Description
DFN-6D	QFN-16C	SOT-25		
1	6, 8, 16	N/A	NC	No connection. Not internally connected. Can left floating or connected to GND.
2	7	1	EN	Enable Control Input, active high.
3	9, 10, 11, 12	4	IN	Input Supply Voltage Pin. Bypass this pin with a capacitor as close to the device as possible.
4	13, 14, 15	3	SW	Switch Node Connection to Inductor.
5	1, 2, 3, 5	2	GND	Ground. Tie directly to ground plane.
6	4	5	FB/OUT	Output voltage Feedback input.

## ■ Ordering Information

**AME5259A - x x x XXX**



Pin Configuration & Special Feature	Package Type	Number of Pins	Output Voltage
A (DFN-6D) 1. NC 2. EN 3. IN 4. SW 5. GND 6. FB/OUT	E: SOT-2X V: DFN W: QFN	E: 16 V: 5 Y: 6	100: 1.0V 120: 1.2V 150: 1.5V 160: 1.6V 180: 1.8V 250: 2.5V 330: 3.3V
A (QFN-16C) 1. GND 2. GND 3. GND 4. FB/OUT 5. GND 6. NC 7. EN 8. NC 9. IN 10. IN 11. IN 12. IN 13. SW 14. SW 15. SW 16. NC			ADJ: Adjustable
A (SOT-25) 1. EN 2. GND 3. SW 4. IN 5. FB/OUT			



AME5259A

## 1.2A, 1.5MHz Synchronous Step-Down Converter

### ■ Absolute Maximum Ratings

Parameter	Maximum	Unit
Input Supply Voltage	-0.3 to 6.5	V
EN, V <sub>OUT</sub> Voltage	-0.3 to V <sub>IN</sub>	
SW Voltage	-0.3 to V <sub>IN</sub>	
ESD Classification	HBM	2 kV
	MM	200 V

### ■ Recommended Operating Conditions

Parameter	Symbol	Rating	Unit
Supply Voltage Voltage	V <sub>IN</sub>	2.5 to 5.5	V
Ambient Temperature Range	T <sub>A</sub>	-40 to +85	°C
Junction Temperature Range	T <sub>J</sub>	-40 to +125	°C



AME5259A

## 1.2A, 1.5MHz Synchronous Step-Down Converter

### ■ Thermal Information

Parameter	Package	Die Attach	Symbol	Maximum	Unit
Thermal Resistance* (Junction to Case)	DFN-6D	Conductive Epoxy	$\theta_{JC}$	85	°C / W
Thermal Resistance (Junction to Ambient)			$\theta_{JA}$	160	
Internal Power Dissipation			$P_D$	625	mW
Thermal Resistance* (Junction to Case)	QFN-16C	Conductive Epoxy	$\theta_{JC}$	67	°C / W
Thermal Resistance (Junction to Ambient)			$\theta_{JA}$	149	
Internal Power Dissipation			$P_D$	670	mW
Thermal Resistance* (Junction to Case)	SOT-25	Conductive Epoxy	$\theta_{JC}$	81	°C / W
Thermal Resistance (Junction to Ambient)			$\theta_{JA}$	260	
Internal Power Dissipation			$P_D$	400	mW
Solder Iron (10Sec)**				350	°C

\* Measure  $\theta_{JC}$  on backside center of Exposed Pad.

\*\* MIL-STD-202G 210F

## ■ Electrical Specifications

$V_{IN}=3.6V$ ,  $V_{OUT}=2.5V$ ,  $V_{FB}=0.6V$ ,  $L=2.2\mu H$ ,  $C_{IN}=4.7\mu F$ ,  $C_{OUT}=10\mu F$ ,  $T_A=25^\circ C$ ,  $I_{MAX}=1.2A$  unless otherwise specified.

Parameter	Symbol	Test Condition	Min	Typ	Max	Units
Input voltage	$V_{IN}$		2.5		5.5	V
Output Voltage Accuracy	$\Delta V_{OUT}$	$V_{IN}=2.5$ to $5.5V$ , in PWM mode For Fixed Output Voltage	-3		3	%
Adjustable Output Range	$V_{out}$		$V_{FB}$		$V_{IN}-0.2$	V
Feedback Voltage	$V_{FB}$	For Adjustable Output Voltage	0.588	0.6	0.612	V
Feedback Pin Bias Current	$I_{FB}$	$V_{FB}=V_{IN}$	-50		50	nA
Quiescent Current	$I_Q$	$I_{OUT}=0mA$ , $V_{FB}=1V$		0.4	0.5	mA
Shutdown Current	$I_{SHDN}$	$V_{EN}=GND$		0.1	1	$\mu A$
Switch Frequency	$f_{osc}$		1.2	1.5	1.8	MHz
High-side Switch On-Resistance	$R_{DS,ON, LHI}$	$I_{SW}=200mA$ , $V_{IN}=3.6V$		0.28		$\Omega$
Low-side Switch On-Resistance	$R_{DS,ON, LO}$	$I_{SW}=200mA$ , $V_{IN}=3.6V$		0.25		$\Omega$
Switch Current Limit	$I_{SW,CL}$	$V_{IN}=2.5$ to $5.5V$		1.6		A
EN High (Enabled the Device)	$V_{EN,HI}$	$V_{IN}=2.5$ to $5.5V$	1.5			V
EN Low (Shutdown the Device)	$V_{EN,LO}$	$V_{IN}=2.5$ to $5.5V$			0.4	V
Input Undervoltage Lockout	$V_{UVLO}$	rising edge		1.8		V
Input Undervoltage Lockout Hysteresis	$V_{UVLO,HYST}$			0.1		V
Thermal Shutdown Temperature	OTP	Shutdown, temperature increasing		160		$^\circ C$
Maximum Duty Cycle	$D_{MAX}$		100			%
SW Leakage Current		$EN=0V$ , $V_{IN}=5.0V$ $V_{SW}=0V$ or $5.0V$	-1		1	$\mu A$



AME

AME5259A

## 1.2A, 1.5MHz Synchronous Step-Down Converter

### ■ Detailed Description

#### Main Control Loop

AME5259A uses a constant frequency, current mode step-down architecture. Both the main (P-channel MOSFET) and synchronous (N-channel MOSFET) switches are internal. During normal operation, the internal top power MOSFET is turned on each cycle when the oscillator sets the RS latch, and turned off when the current comparator resets the RS latch. While the top MOSFET is off, the bottom MOSFET is turned on until either the inductor current starts to reverse as indicated by the current reversal comparator IRCPMP.

#### Pulse Skipping Mode Operation

At light loads, the inductor current may reach zero or reverse on each pulse. The bottom MOSFET is turned off by the current reversal comparator, IRCPMP, and the switch voltage will ring. This is discontinuous mode operation, and is normal behavior for the switching regulator.

#### Short-Circuit Protection

When the output is shorted to ground, the frequency of the oscillator is reduced to about 180KHz. This frequency foldback ensures that the inductor current has more time to decay, thereby preventing runaway. The oscillator's frequency will progressively increase to 1.5MHz when  $V_{FB}$  or  $V_{OUT}$  rises above 0V.

#### Dropout Operation

As the input supply voltage decreases to a value approaching the output voltage, the duty cycle increases toward the maximum on-time. Further reduction of the supply voltage forces the main switch to remain on for more than one cycle until it reaches 100% duty cycle. The output voltage will then be determined by the input voltage minus the voltage drop across the P-channel MOSFET and the inductor.

### ■ Application Information

The basic AME5259A application circuit is shown in Typical Application Circuit. External component selection is determined by the maximum load current and begins with the selection of the inductor value and followed by  $C_{IN}$  and  $C_{OUT}$ .

#### Inductor Selection

For a given input and output voltage, the inductor value and operating frequency determine the ripple current. The ripple current  $\Delta I_L$  increases with higher  $V_{IN}$  and decreases with higher inductance.

$$\Delta I_L = \frac{1}{f \times L} \times V_{OUT} \left(1 - \frac{V_{OUT}}{V_{IN}}\right)$$

A reasonable starting point for setting ripple current is  $\Delta I_L = 0.4(I_{max})$ . The DC current rating of the inductor should be at least equal to the maximum load current plus half the ripple current to prevent core saturation. For better efficiency, choose a low DC-resistance inductor.

#### $C_{IN}$ and $C_{OUT}$ Selection

The input capacitance,  $C_{IN}$  is needed to filter the trapezoidal current at the source of the top MOSFET. To prevent large voltage transients, a low ESR input capacitor sized for the maximum RMS current must be used. The maximum RMS capacitor current is given by:

$$I_{RMS} = I_{OUT(MAX)} \times \frac{V_{OUT}}{V_{IN}} \times \sqrt{\frac{V_{IN}}{V_{OUT}} - 1}$$

This formula has a maximum at  $V_{IN}=2V_{OUT}$ , where  $I_{RMS}=I_{OUT}/2$ . This simple worst-case condition is commonly used for design because even significant deviations do not offer much relief. Note that the capacitor manufacturer ripple current ratings are often based on 2000 hours of life. This makes it advisable to further derate the capacitor, or choose a capacitor rated at a higher temperature than required.

The selection of  $C_{OUT}$  is determined by the effective series resistance (ESR) that is required to minimize voltage ripple and load step transients. The output ripple,  $V_{OUT}$ , is determined by:

$$\Delta V_{OUT} \cong \Delta I_L \left( ESR + \frac{1}{8fC_{OUT}} \right)$$

## ■ Detailed Description

### Using Ceramic Input and Output Capacitors

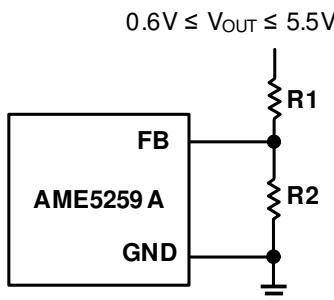
Higher values, lower cost ceramic capacitors are now becoming available in smaller case sizes. Their high ripple current, high voltage rating and low ESR make them ideal for switching regulator applications. However, care must be taken when these capacitors are used at the input and output. When a ceramic capacitor is used at the input and the power is supplied by a wall adapter through long wires, a load step at the output can induce ringing at the input,  $V_{IN}$ . At best, this ringing can couple to the output and be mistaken as loop instability. At worst, a sudden inrush of current through the long wires can potentially cause a voltage spike at  $V_{IN}$  large enough to damage the part.

### Output Voltage Programming

The output voltage is set by an external resistive divider according to the following equation :

$$V_{OUT} = V_{REF} \cdot \left(1 + \frac{R_1}{R_2}\right)$$

Where VREF equals to 0.6V typical. The resistive divider allows the FB pin to sense a fraction of the output voltage as shown in Figure 4.



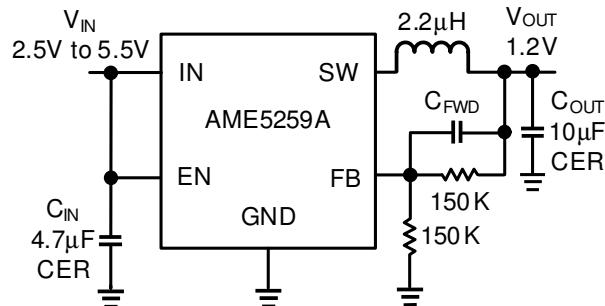
**Figure 4. Setting the AME5259A Output Voltage**

### Thermal Considerations

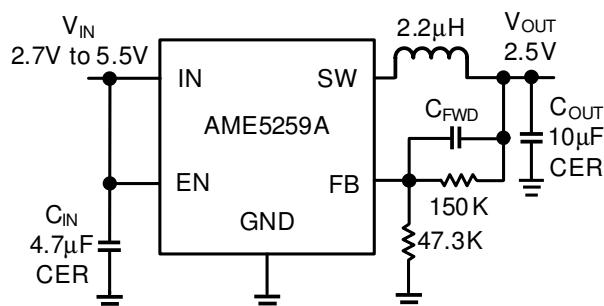
In most applications the AME5259A does not dissipate much heat due to its high efficiency. But, in applications where the AME5259A is running at high ambient temperature with low supply voltage and high duty cycles, such as in dropout, the heat dissipated may exceed the maximum junction temperature of the part. If the junction temperature reaches approximately 160°C, both power switches will be turned off and the SW node will become high impedance. To avoid the AME5259A from exceeding the maximum junction temperature, the user will need to do some thermal analysis. The goal of the thermal analysis is to determine whether the power dissipated exceeds the maximum junction temperature of the part. The temperature rise is given by:

$$T_R = (PD)(\theta_{JA})$$

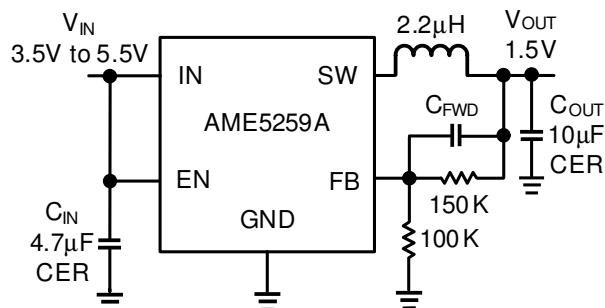
Where PD is the power dissipated by the regulator and  $\theta_{JA}$  is the thermal resistance from the junction of the die to the ambient temperature.



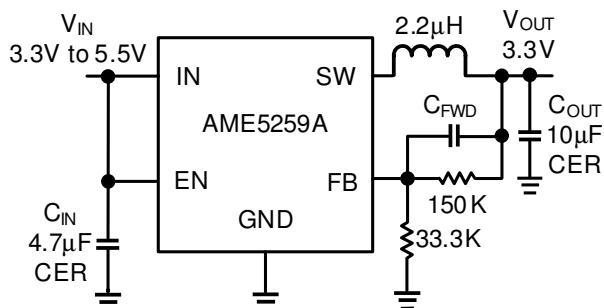
**Figure 5. 1.2V Step-Down Regulator**  
 $C_{FWD}$ : 22pF~220 pF



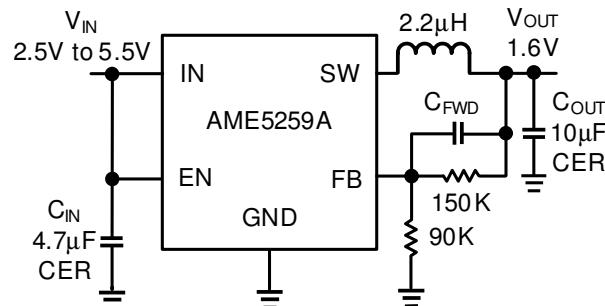
**Figure 8. 2.5V Step-Down Regulator**  
 $C_{FWD}$ : 22pF~220 pF



**Figure 6. 1.5V Step-Down Regulator**  
 $C_{FWD}$ : 22pF~220 pF



**Figure 9. 3.3V Step-Down Regulator**  
 $C_{FWD}$ : 22pF~220 pF



**Figure 7. 1.6V Step-Down Regulator**  
 $C_{FWD}$ : 22pF~220 pF

### PC Board Layout Checklist

When laying out the printed circuit board, the following checklist should be used to ensure proper operation of the AME5259A. These items are also illustrated graphically in Figures 10 and Figures 11 . Check the following in your layout:

1. The power traces, consisting of the GND trace, the SW trace and the  $V_{IN}$  trace should be kept short, direct and wide.
2. Does the  $V_{FB}$  pin connect directly to the feedback resistors? The resistive divider  $R2/R1$  must be connected between the (+) plate of  $C_{OUT}$  and ground.
3. Does the (+) plate of  $C_{IN}$  connect to  $V_{IN}$  as closely as possible? This capacitor provides the AC current to the internal power MOSFETs.
4. Keep the switching node, SW, away from the sensitive  $V_{FB}$  node.
5. Keep the (-) plates of  $C_{IN}$  and  $C_{OUT}$  as close as possible.

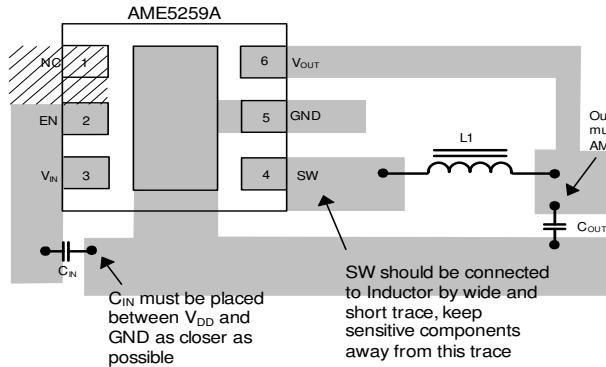
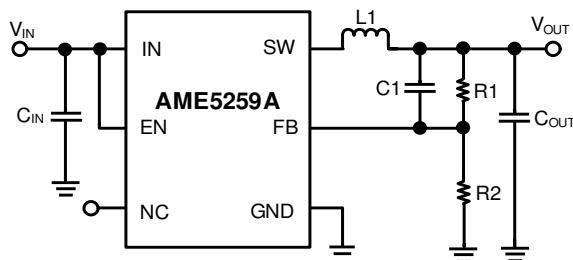


Figure 10. AME5259A Adjustable Voltage Regulator Layout Diagram

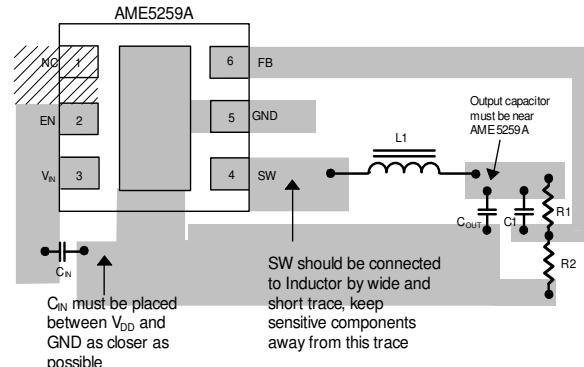
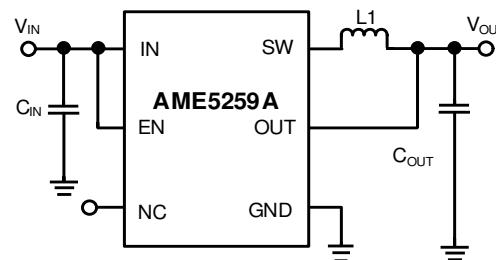


Figure 11. AME5259A Fixed Voltage Regulator Layout Diagram

**AME5259A****1.2A, 1.5MHz Synchronous  
Step-Down Converter****■ Application Information**

Supplier	Inductance ( $\mu$ H)	Current Rating (mA)	DCR (m $\Omega$ )	Dimensions (mm)	Series
TAIYO YUDEN	2.2	1480	60	3.00 x 3.00 x 1.50	NR3015
GOTREND	2.2	1500	58	3.85 x 3.85 x 1.80	GTSD32
Sumida	2.2	1500	75	4.50 x 3.20 x 1.55	CDRH2D14

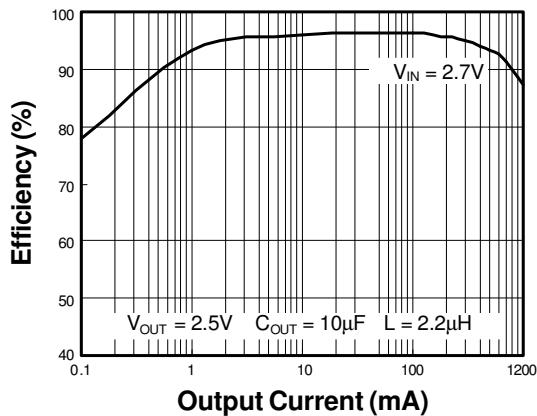
**Table 1. Recommended Inductors**

Supplier	Capacitance ( $\mu$ F)	Package	Part Number
TDK	4.7	603	C1608JB0J475M
MURATA	4.7	603	GRM188R60J475KE19
TAIYO YUDEN	4.7	603	JMK107BJ475RA
TAIYO YUDEN	10	603	JMK107BJ475RA
TDK	10	805	C2012JB0J106M
MURATA	10	805	GRM219R60J106ME19
MURATA	10	805	GRM219R60J106ME19
TAIYO YUDEN	10	805	JMK212BJ106RD

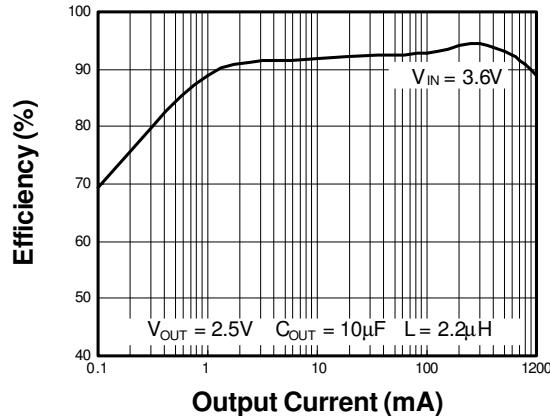
**Table 2. Recommended Capacitors for  $C_{IN}$  and  $C_{OUT}$**

## ■ Characterization Curve

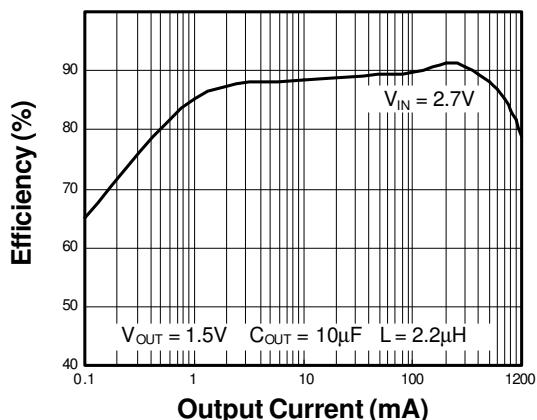
**Efficiency vs. Output Current**



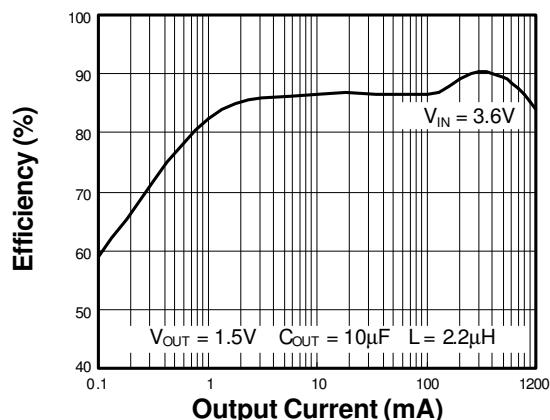
**Efficiency vs. Output Current**



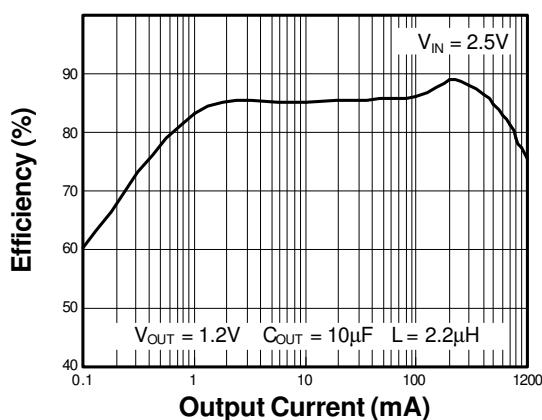
**Efficiency vs. Output Current**



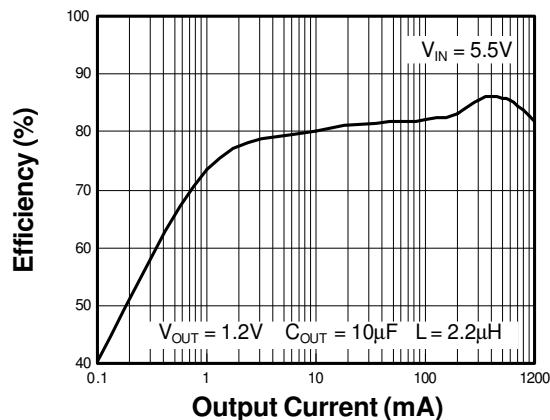
**Efficiency vs. Output Current**



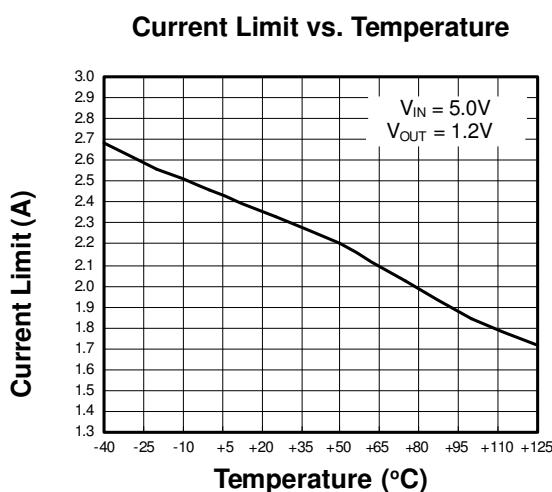
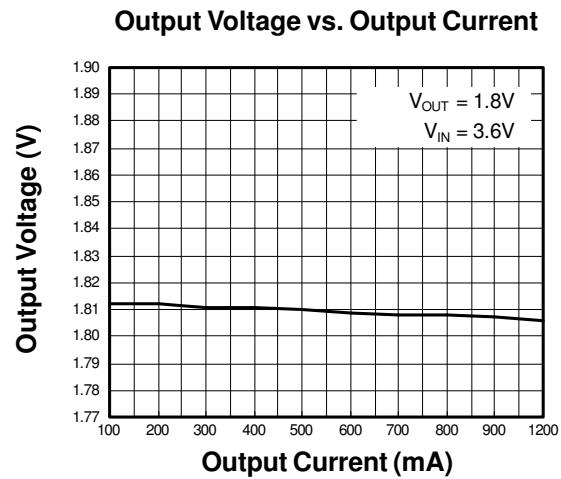
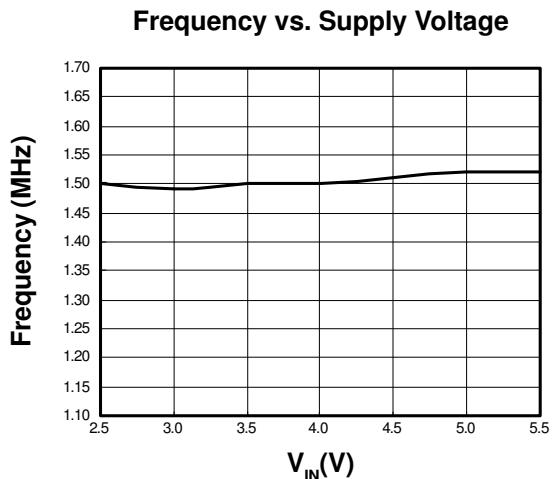
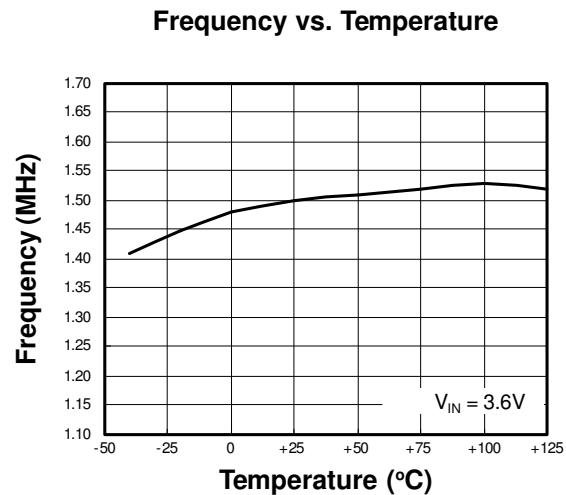
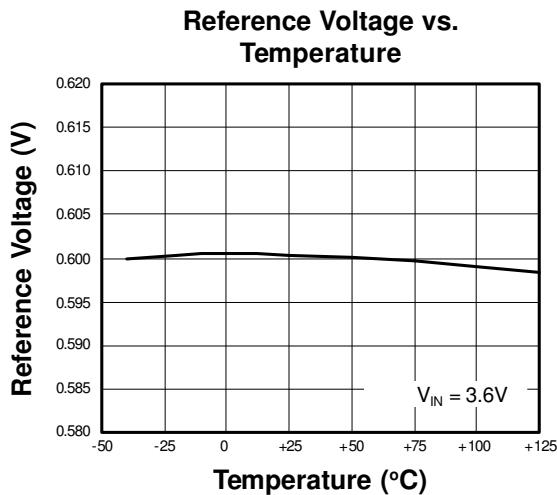
**Efficiency vs. Output Current**



**Efficiency vs. Output Current**

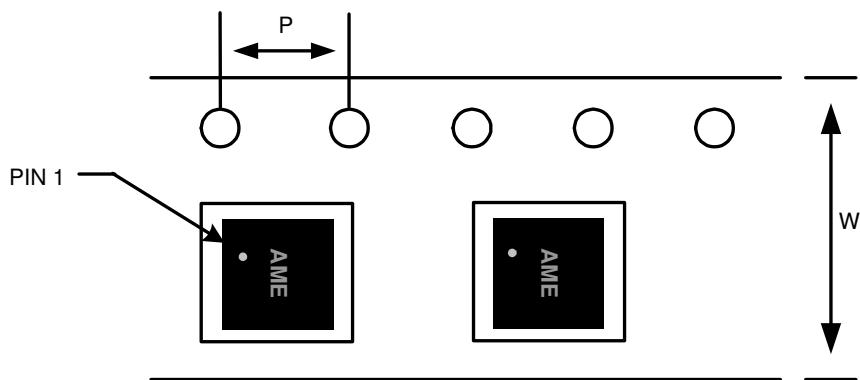


## ■ Characterization Curve (Contd.)



## ■ Tape and Reel Dimension

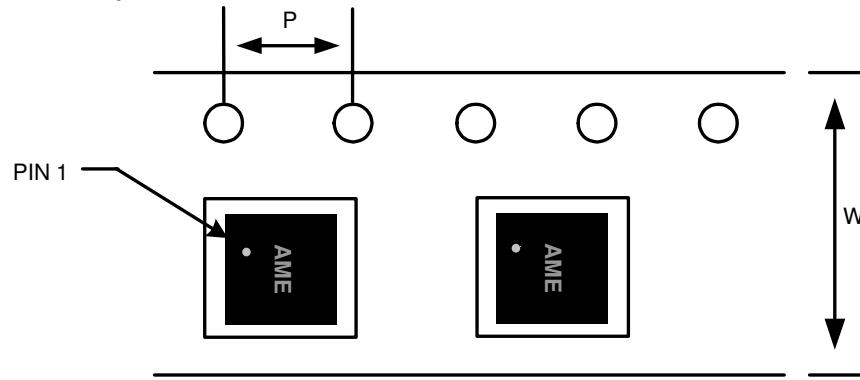
**DFN-6D**  
(2mmx2mmx0.75mm)



Carrier Tape, Number of Components Per Reel and Reel Size

Package	Carrier Width (W)	Pitch (P)	Part Per Full Reel	Reel Size
DFN-6D (2x2x0.75mm)	8.0±0.1 mm	4.0±0.1 mm	3000pcs	180±1 mm

**QFN-16C**  
(3mmx3mmx0.75mm)

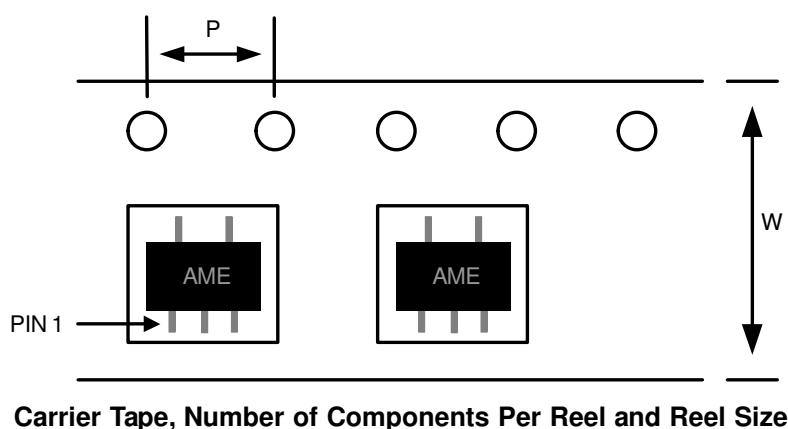


Carrier Tape, Number of Components Per Reel and Reel Size

Package	Carrier Width (W)	Pitch (P)	Part Per Full Reel	Reel Size
QFN-16C (3x3x0.75mm)	12.0±0.1 mm	4.0±0.1 mm	3000pcs	330±1 mm

## ■ Tape and Reel Dimension (Contd.)

SOT-25

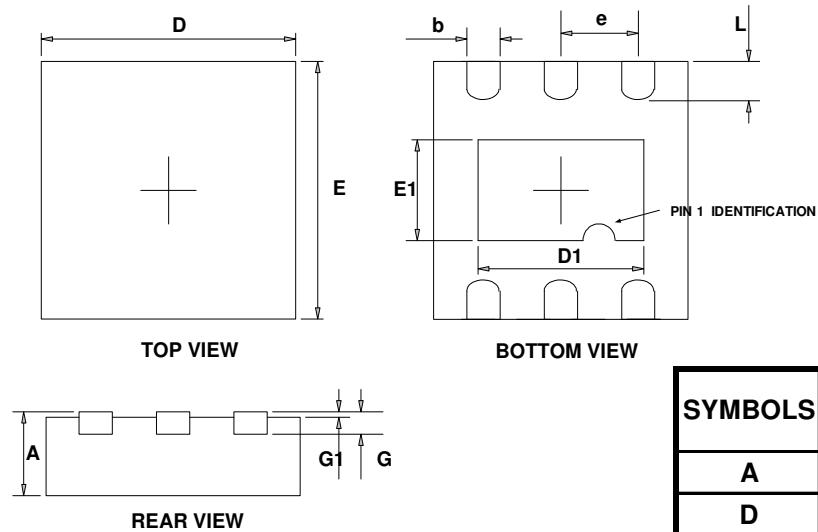


**Carrier Tape, Number of Components Per Reel and Reel Size**

Package	Carrier Width (W)	Pitch (P)	Part Per Full Reel	Reel Size
SOT-25	$8.0 \pm 0.1$ mm	$4.0 \pm 0.1$ mm	3000pcs	$180 \pm 1$ mm

## ■ Package Dimension

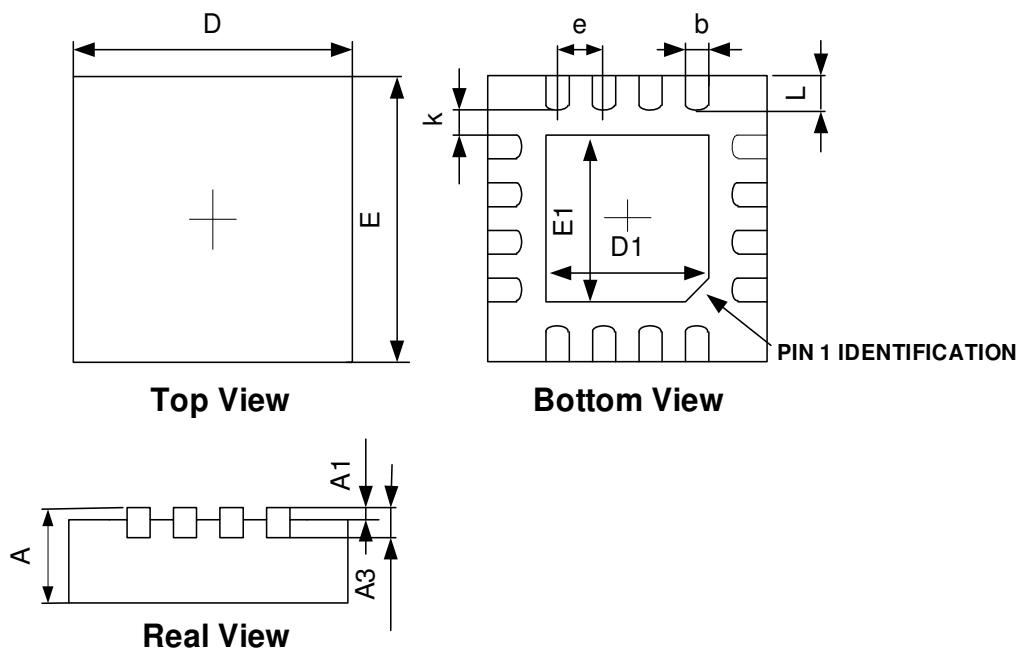
**DFN-6D**  
(2mmx2mmx0.75mm)



SYMBOLS	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	0.700	0.800	0.028	0.031
D	1.900	2.100	0.075	0.083
E	1.900	2.100	0.075	0.083
e	0.650 TYP		0.026 TYP	
D1	1.100	1.650	0.043	0.065
E1	0.600	1.050	0.024	0.041
b	0.180	0.350	0.007	0.014
L	0.200	0.450	0.008	0.018
G	0.178	0.228	0.007	0.009
G1	0.000	0.050	0.000	0.002

## ■ Package Dimension

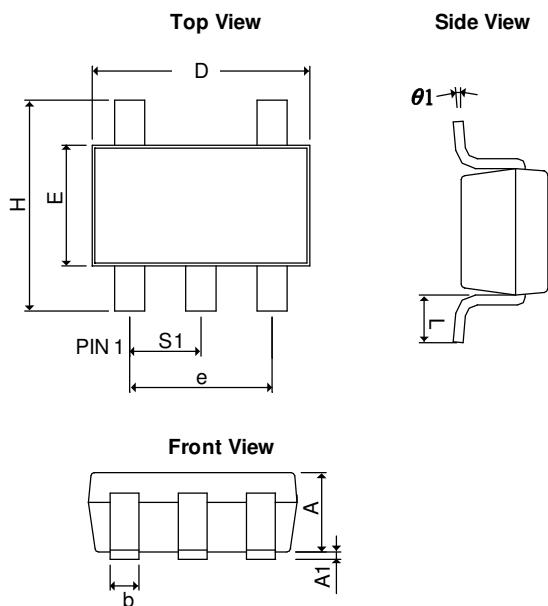
**QFN-16C**  
 (3mmx3mmx0.75mm)



SYMBOLS	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
<b>A</b>	0.700	0.800	0.028	0.031
<b>A1</b>	0.000	0.050	0.000	0.002
<b>A3</b>	0.203REF.		0.008REF.	
<b>D</b>	2.924	3.076	0.115	0.121
<b>E</b>	2.924	3.076	0.115	0.121
<b>D1</b>	1.600	1.800	0.063	0.071
<b>E1</b>	1.600	1.800	0.063	0.071
<b>k</b>	0.200MIN.		0.008MIN.	
<b>b</b>	0.180	0.280	0.007	0.011
<b>e</b>	0.500TYP.		0.020TYP.	
<b>L</b>	0.324	0.476	0.013	0.019

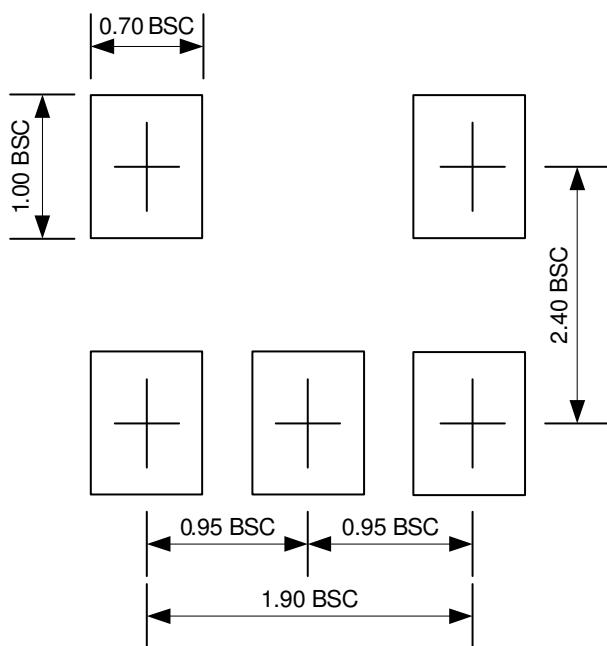
## ■ Package Dimension

SOT-25



SYMBOLS	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	0.90	1.30	0.0354	0.0512
$A_1$	0.00	0.15	0.0000	0.0059
b	0.30	0.55	0.0118	0.0217
D	2.70	3.10	0.1063	0.1220
E	1.40	1.80	0.0551	0.0709
e	1.90 BSC		0.0748 BSC	
H	2.60	3.00	0.1024	0.1181
L	0.37 BSC		0.0146 BSC	
$\theta_1$	$0^\circ$	$10^\circ$	$0^\circ$	$10^\circ$
$S_1$	0.95 BSC		0.0374 BSC	

## ■ Lead Pattern



Note:

1. Lead pattern unit description:  
BSC: Basic. Represents theoretical exact dimension or dimension target.
2. Dimensions in Millimeters.
3. General tolerance  $\pm 0.05\text{mm}$  unless otherwise specified.



**www.ame.com.tw**  
**E-Mail: sales@ame.com.tw**

**Life Support Policy:**

These products of AME, Inc. are not authorized for use as critical components in life-support devices or systems, without the express written approval of the president of AME, Inc.

AME, Inc. reserves the right to make changes in the circuitry and specifications of its devices and advises its customers to obtain the latest version of relevant information.

© AME, Inc. , January 2014  
Document: 1283-DS5259A-A.02

**Corporate Headquarter  
AME, Inc.**

8F, 12, WenHu St., Nei-Hu  
Taipei 114, Taiwan.  
Tel: 886 2 2627-8687  
Fax: 886 2 2659-2989